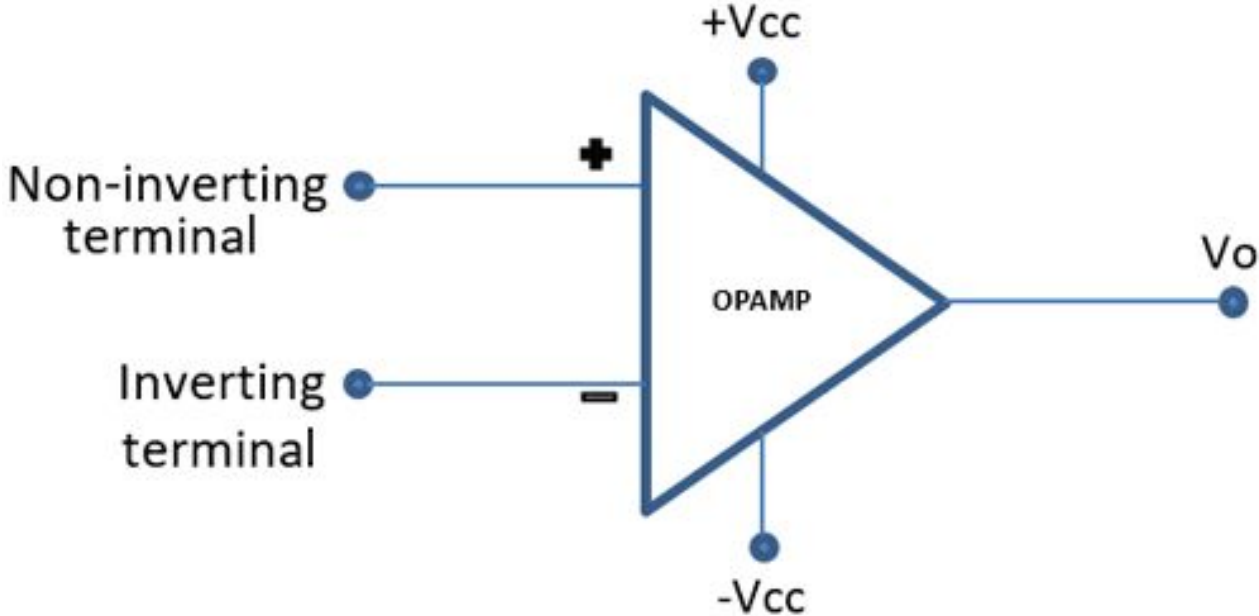


Operational Amplifiers

John R. Leeman

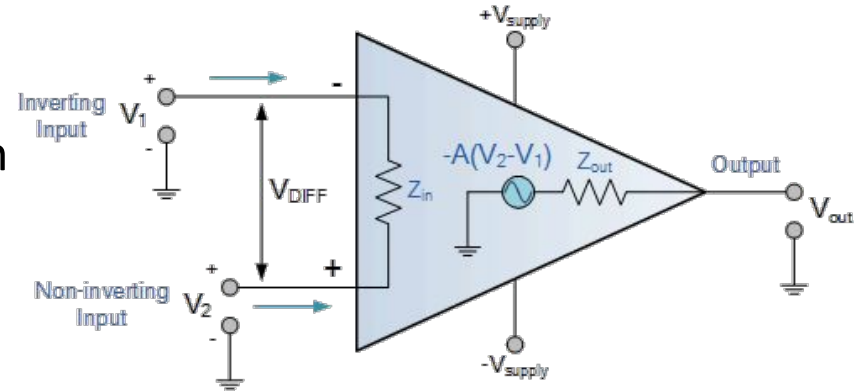
GEARS 2023

Op-Amps have just a few basic terminals, some with other features

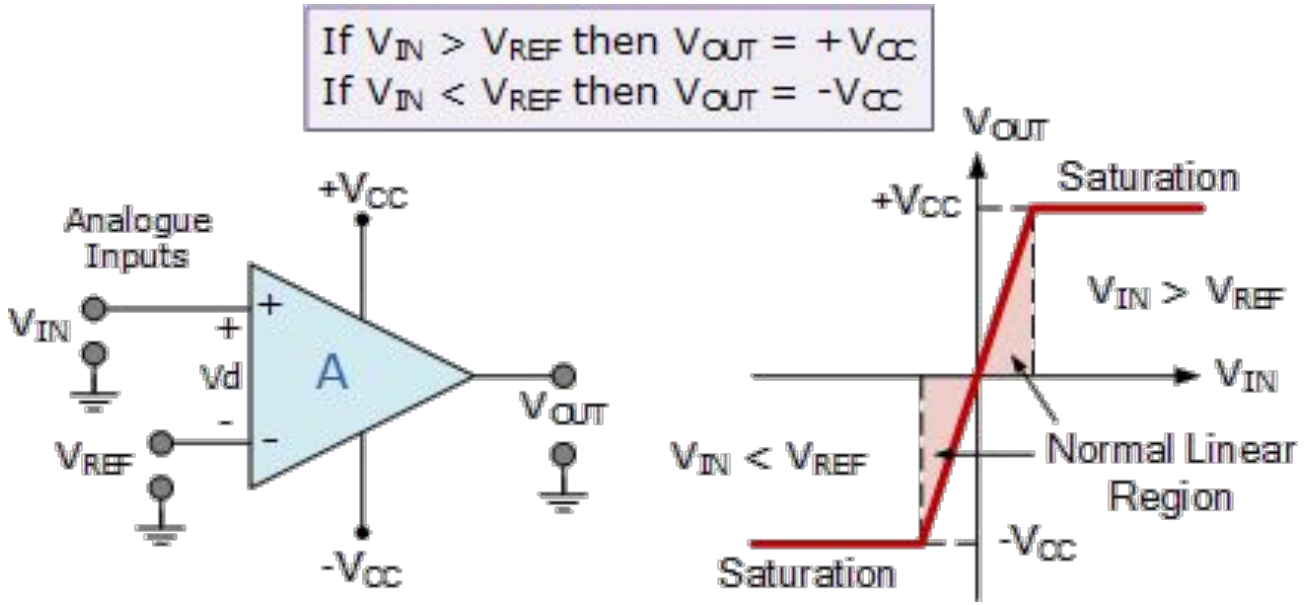


A few rules let us understand ANY Op-Amp circuit

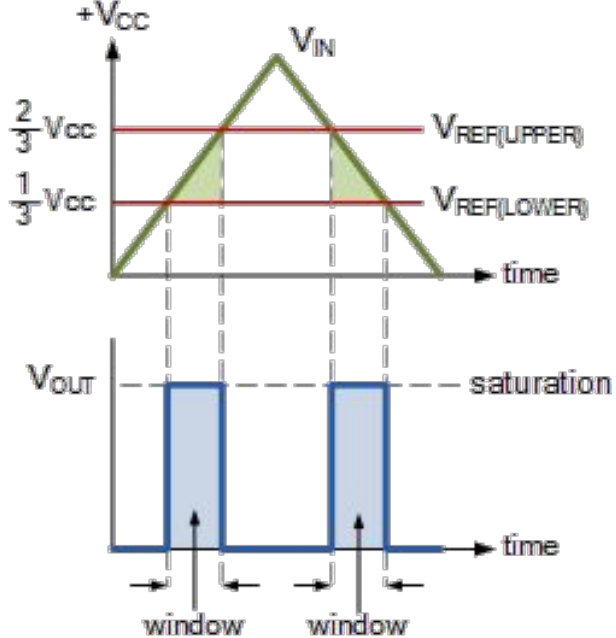
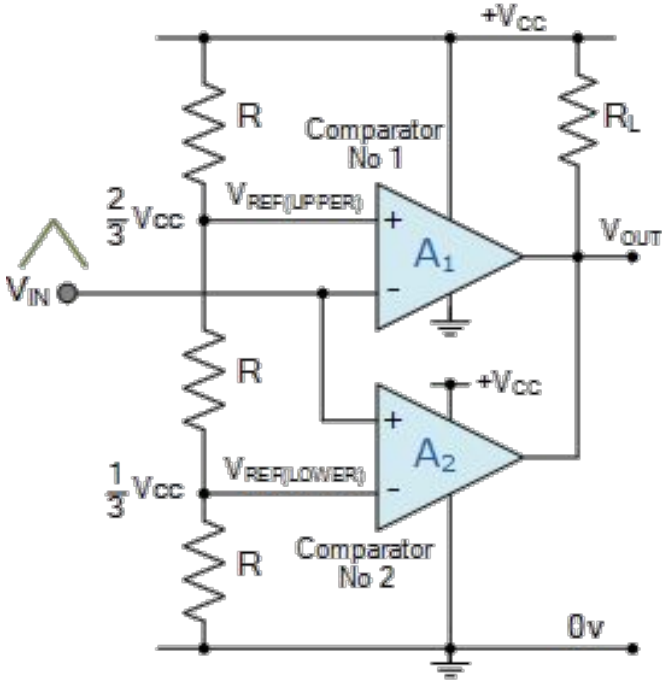
- 1) No current flow into/out of the input terminals (high impedance)
- 2) Op Amp will do whatever is necessary to keep the inputs at the same voltage (when feedback is used)
- 3) The output can source any current (low impedance)
- 4) Open loop gain is large



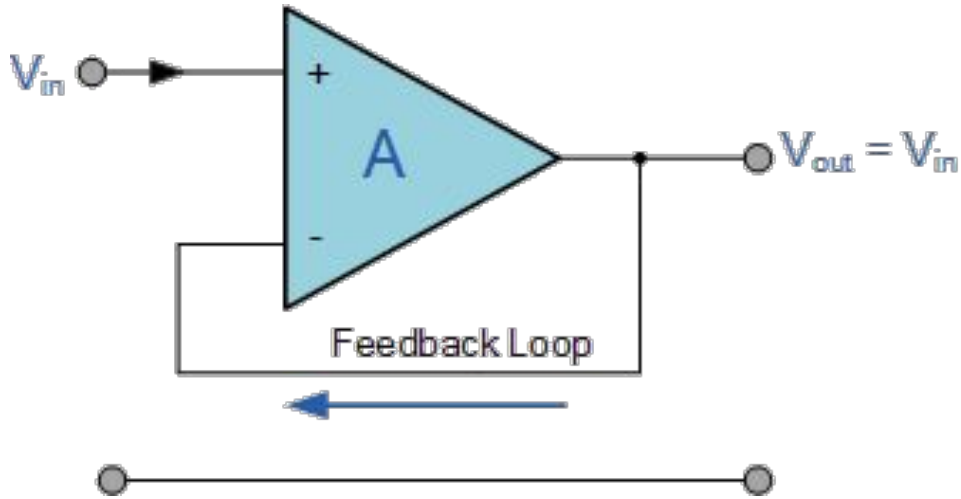
The most straightforward circuit is a comparator



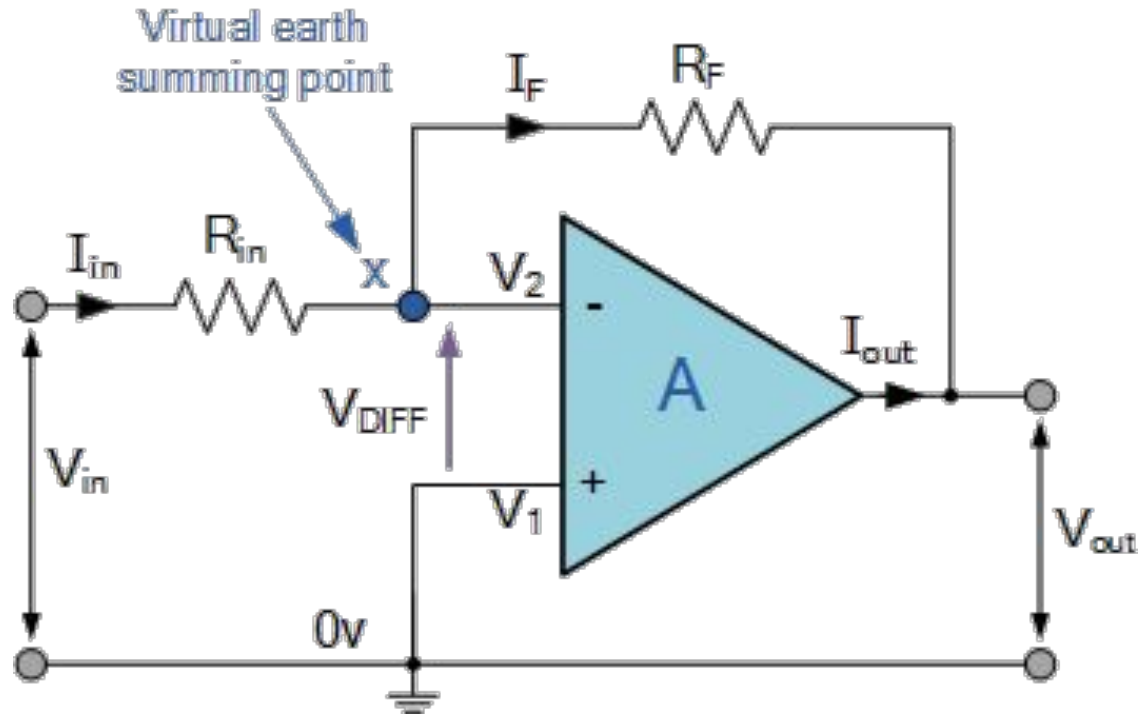
Two comparators can make a window comparator



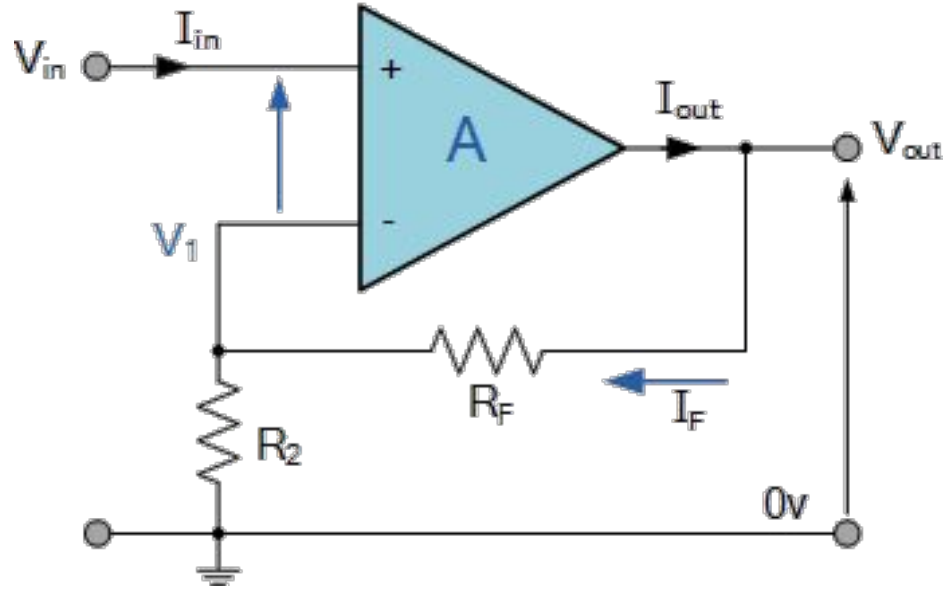
One of the most useful circuits is a voltage follower or buffer



Inverting amplifiers have a simple gain formula, but some caveats



Non-inverting amplifiers are modified voltage followers

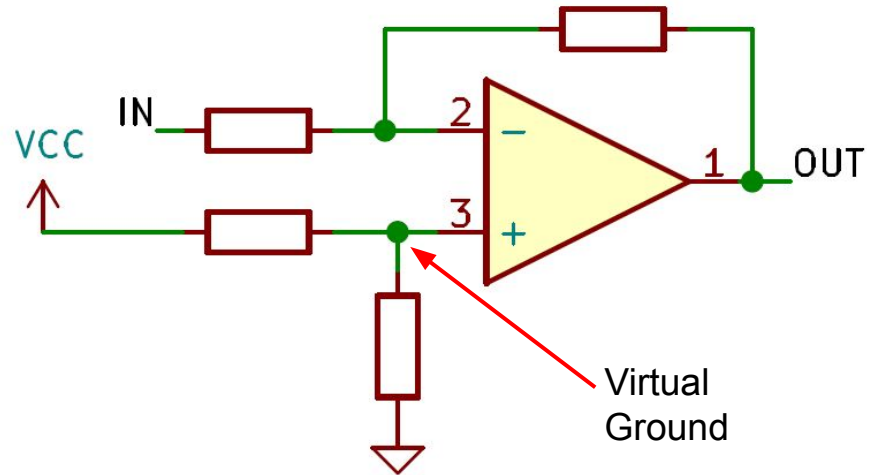


Amplifiers don't have to operate from a bipolar supply



Single-Supply Op Amp Design Techniques

*Application
Report*

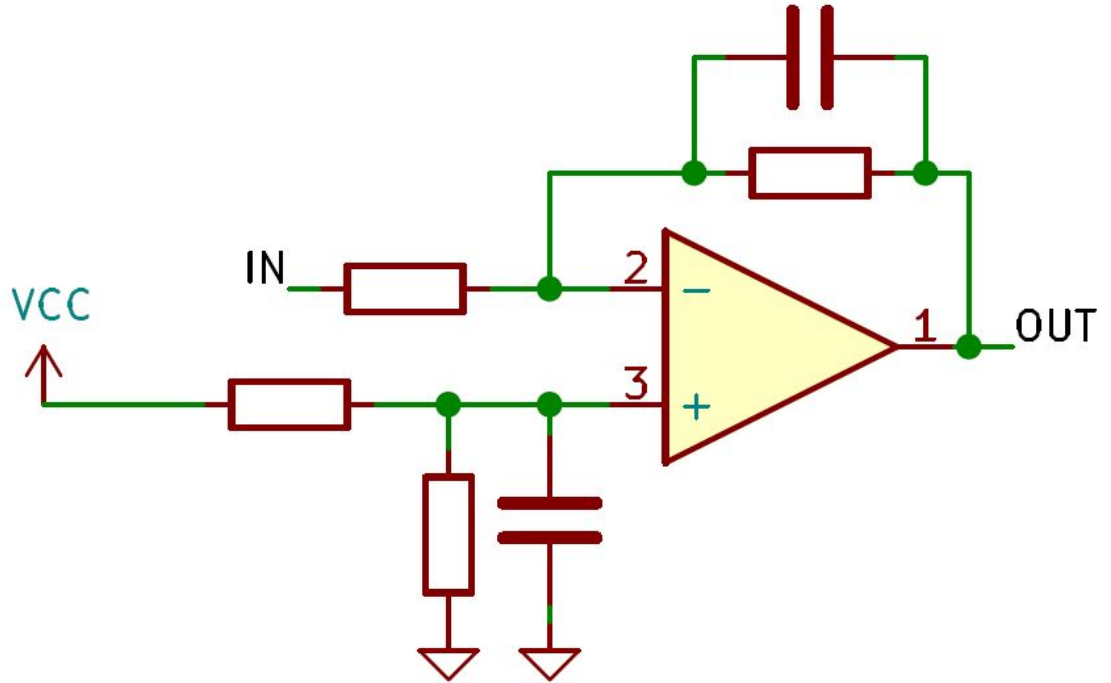


Amplifiers don't have to operate from a bipolar supply

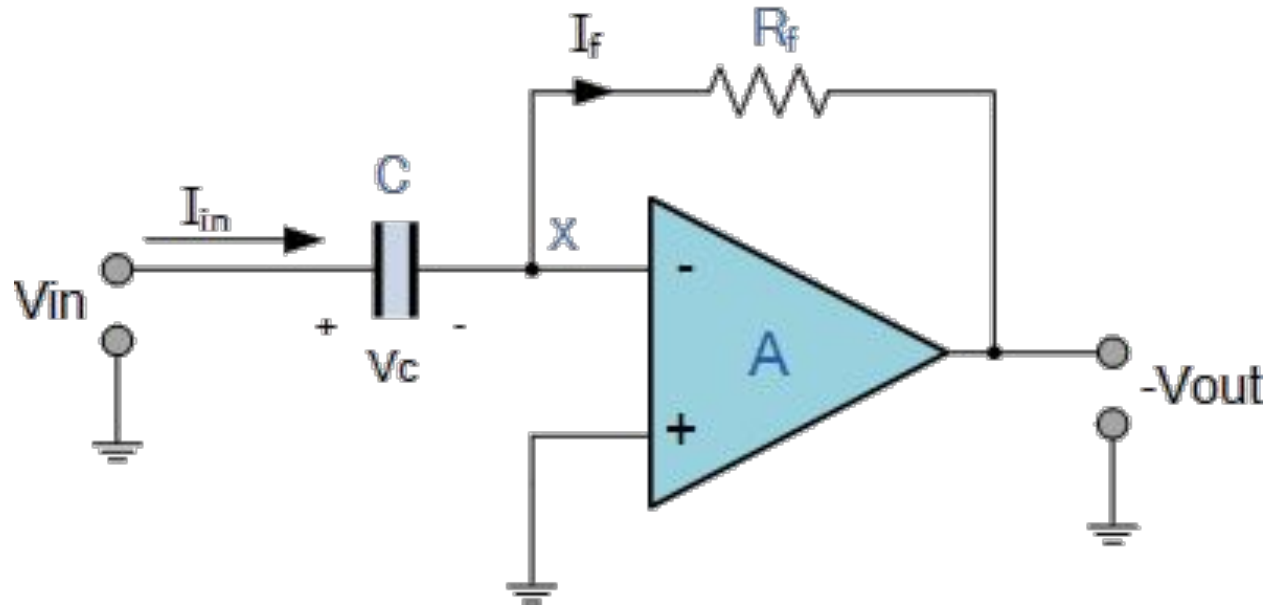


Single-Supply Op Amp Design Techniques

Application Report

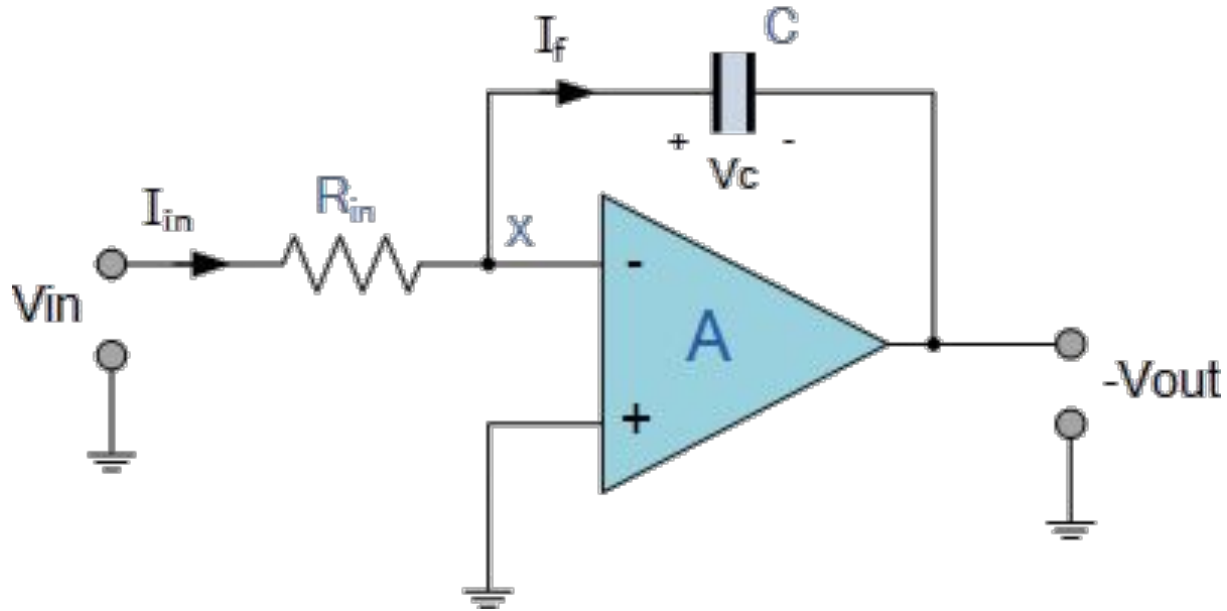


Differentiators output the rate of change of the input



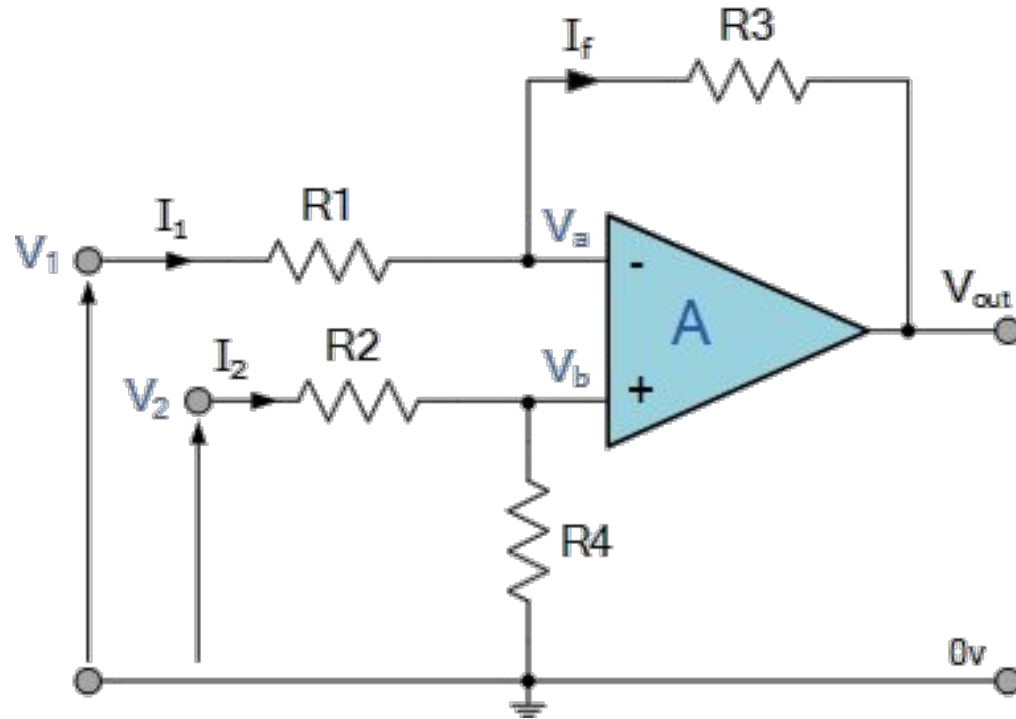
$$V_{OUT} = -R_F C \frac{dV_{IN}}{dt}$$

Integrators integrate the input over a time window

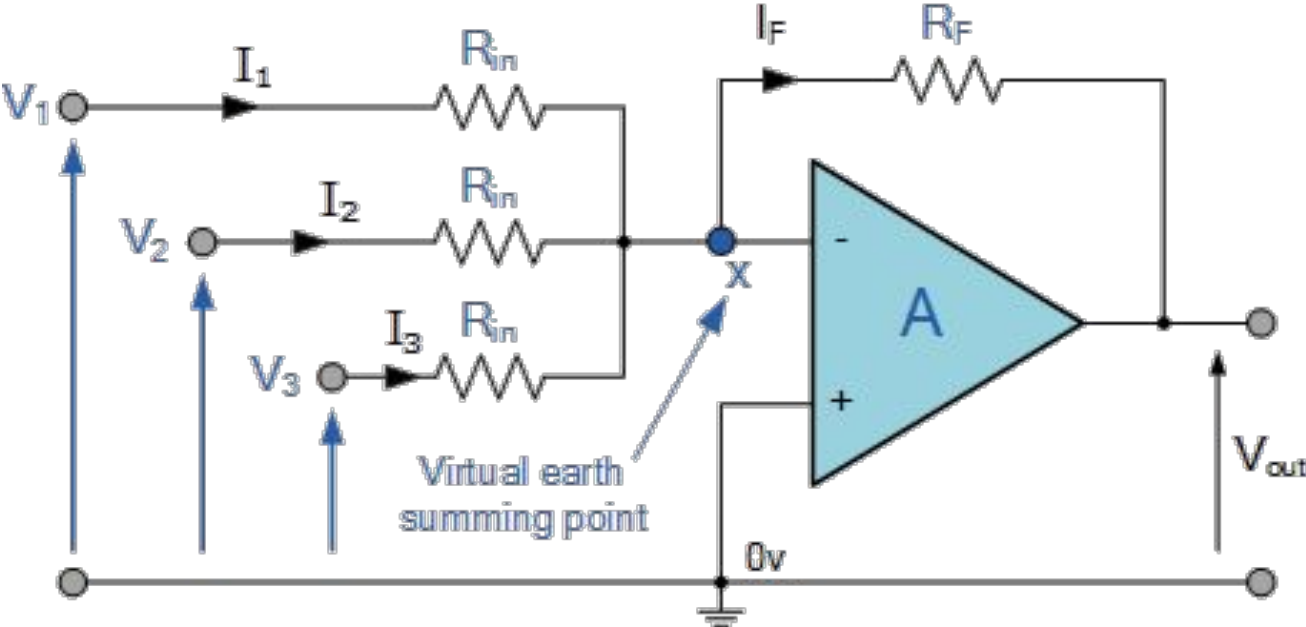


$$V_{out} = -\frac{1}{R_{in} C} \int_0^t V_{in} dt = -\int_0^t V_{in} \frac{dt}{R_{in} \cdot C}$$

Differential amplifiers work with “double ended” signals, but aren’t commonly implemented



Summing amplifiers can be handy as well



$$-V_{out} = \left[\frac{R_F}{R_{in}} V_1 + \frac{R_F}{R_{in}} V_2 + \frac{R_F}{R_{in}} V_3 \right]$$

There are a lot of traps out there that can surprise even experienced electrical engineers



- 1) Input Bias Current
- 2) Input Offset Voltage
- 3) Gain Bandwidth Product
- 4) Others!

Input bias current is current flowing into/out of the inputs

6.5 Electrical Characteristics, LM741⁽¹⁾

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT
Input offset voltage	$R_S \leq 10 \text{ k}\Omega$	$T_A = 25^\circ\text{C}$		1	5	mV
		$T_{AMIN} \leq T_A \leq T_{AMAX}$			6	mV
Input offset voltage adjustment range	$T_A = 25^\circ\text{C}, V_S = \pm 20 \text{ V}$			± 15		mV
Input offset current	$T_A = 25^\circ\text{C}$			20	200	nA
	$T_{AMIN} \leq T_A \leq T_{AMAX}$			85	500	
Input bias current	$T_A = 25^\circ\text{C}$			80	500	nA
	$T_{AMIN} \leq T_A \leq T_{AMAX}$				1.5	μA

Input bias current also depends on everything else

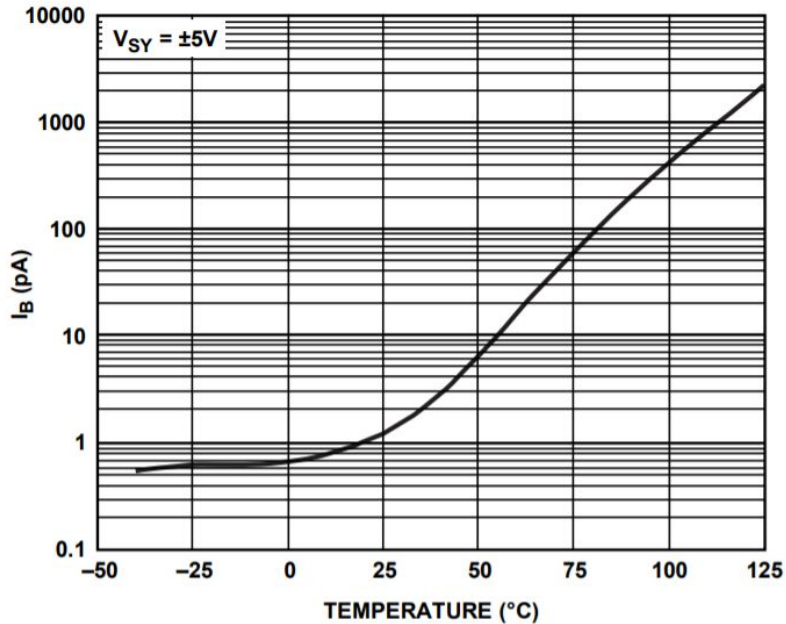


Figure 12. Input Bias Current vs. Temperature

07670-012

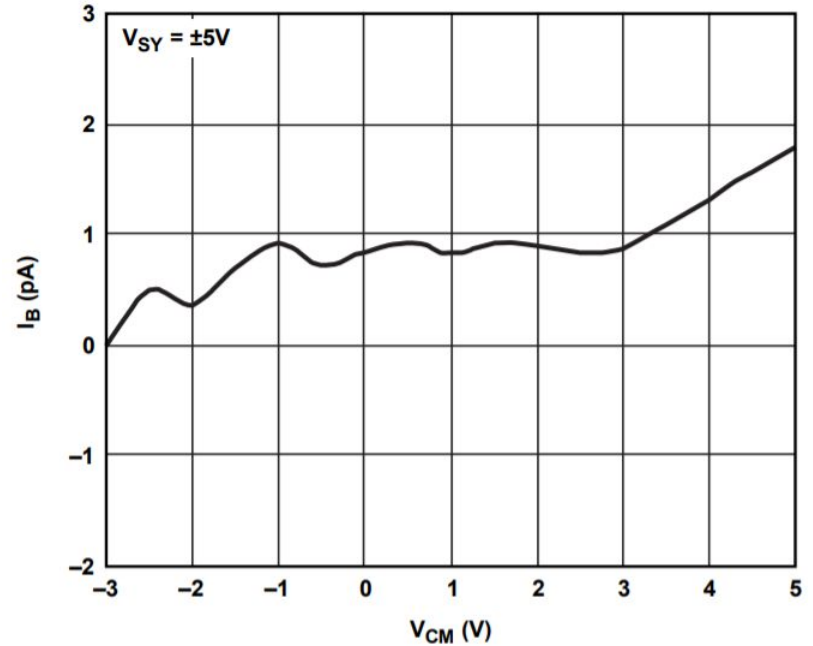
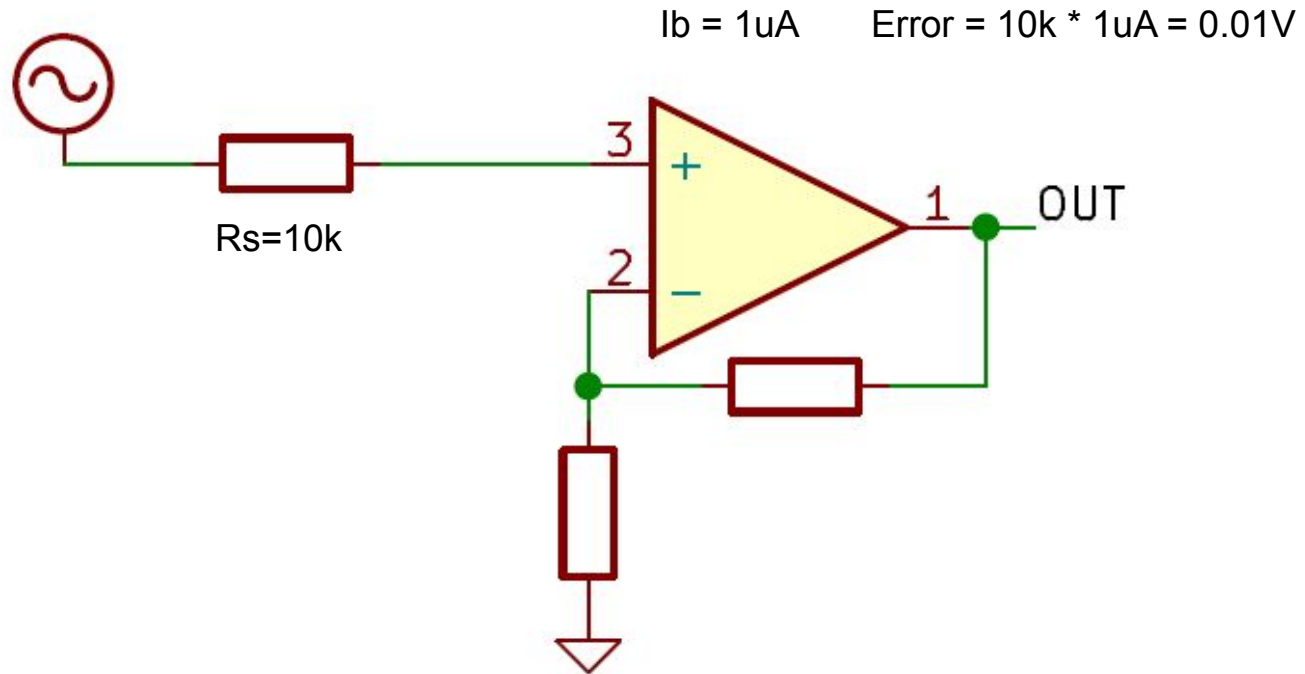


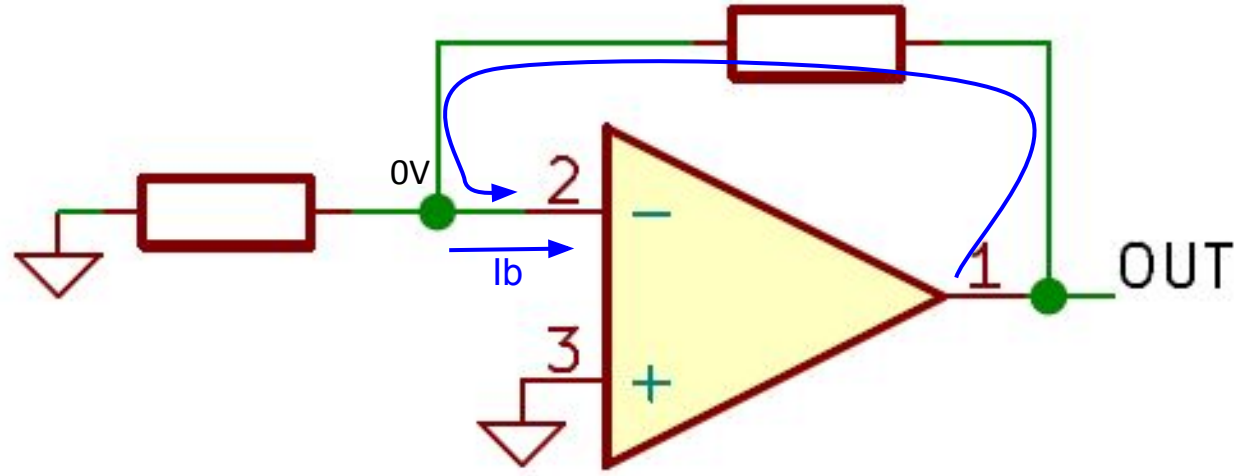
Figure 13. Input Bias Current vs. Common-Mode Voltage

07670-013

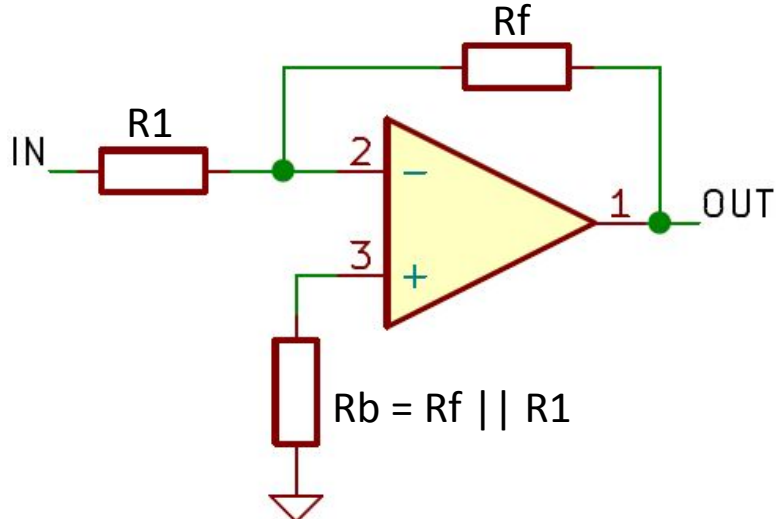
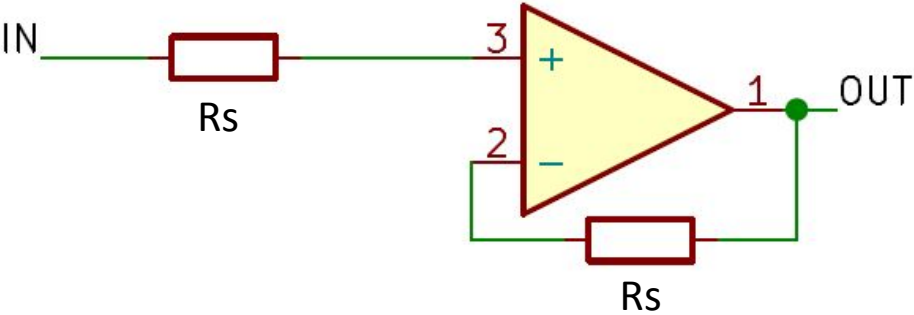
Input Bias Current Problem #1 - Source Impedance



Input Bias Current Problem #2 - Gain Network



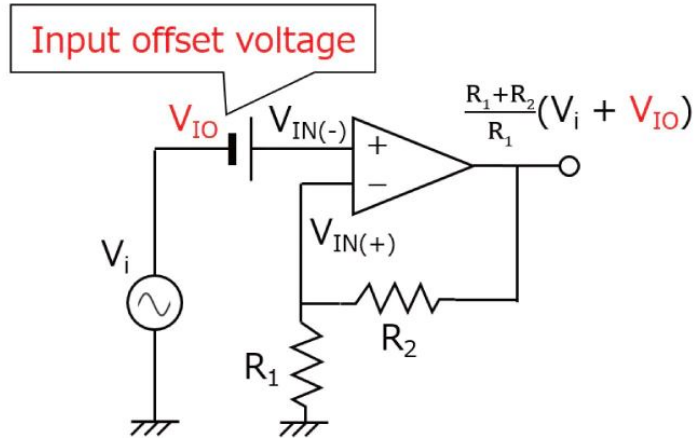
Some solutions



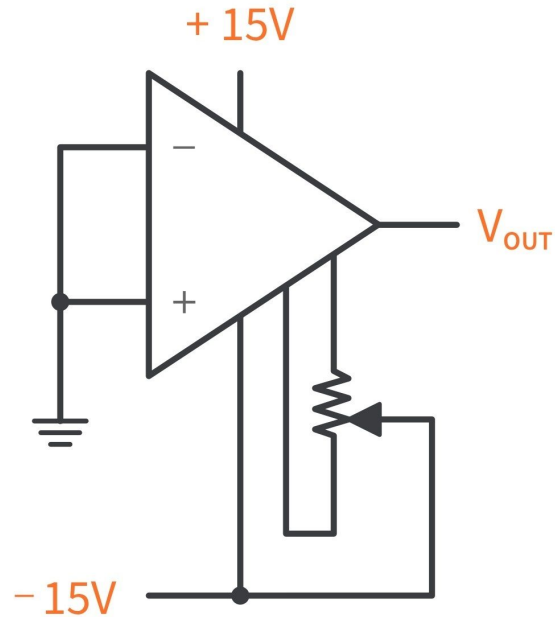
Input offset voltage is another hidden gotcha

Table 2.

Parameter	Symbol	Conditions	Min	Typ	Max	Unit
INPUT CHARACTERISTICS						
Offset Voltage	V_{OS}	$-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$		0.5	1.5	mV
B Grade (ADA4062-2, 8-Lead SOIC Only)						
A Grade				0.75	2.5	mV
Offset Voltage Drift	$\Delta V_{OS}/\Delta T$	$-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$		5	5	$\mu\text{V}/^{\circ}\text{C}$



Vios can be compensated for with null terminals on some amplifiers



$V_{OUT} =$ Potentiometer adjusted so that 0 volts with inputs shorted together

A quick aside to talk about decibels (dBs) as the engineer thinks

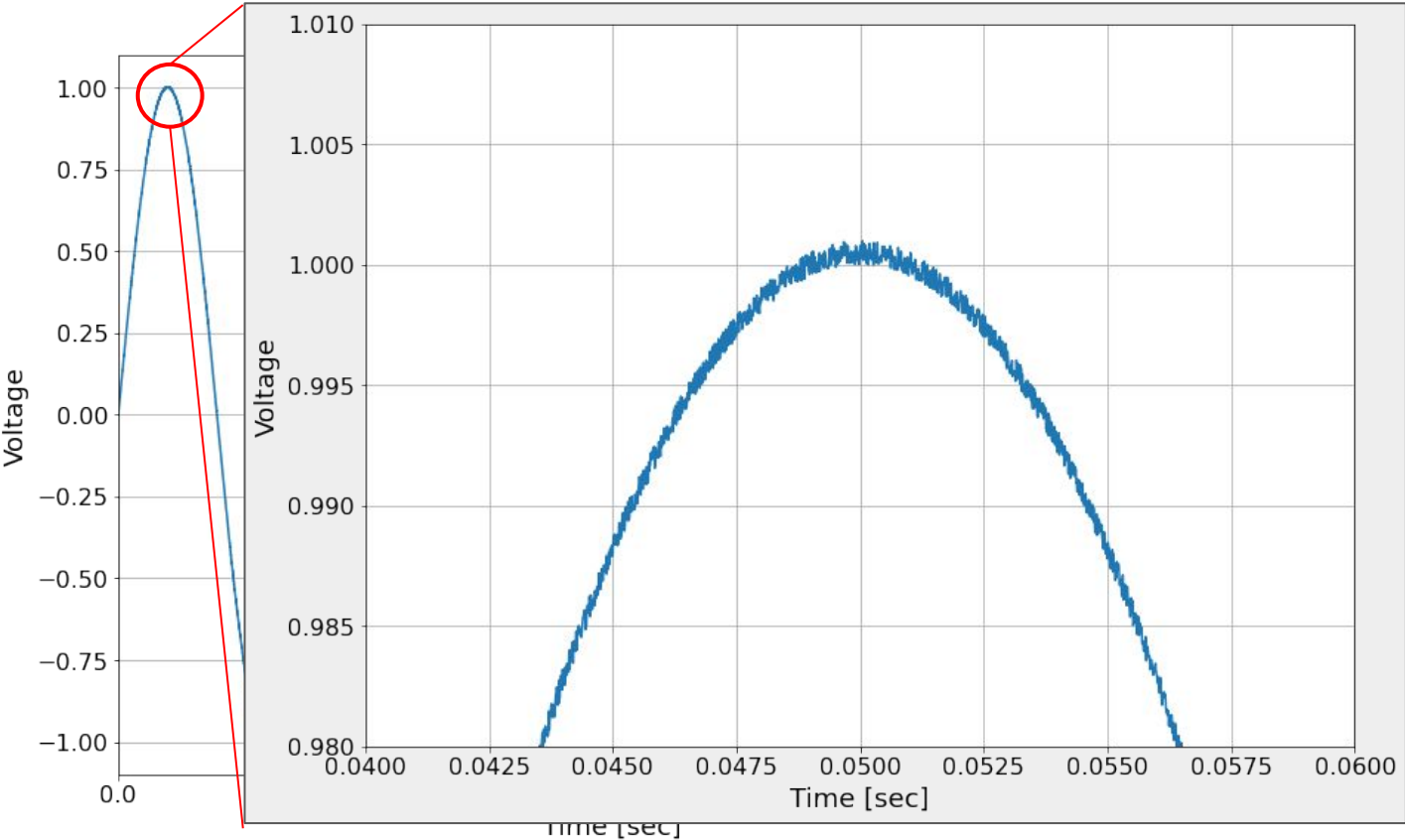


$$\text{dB} = 10 \log_{10} \frac{P_1}{P_2}$$

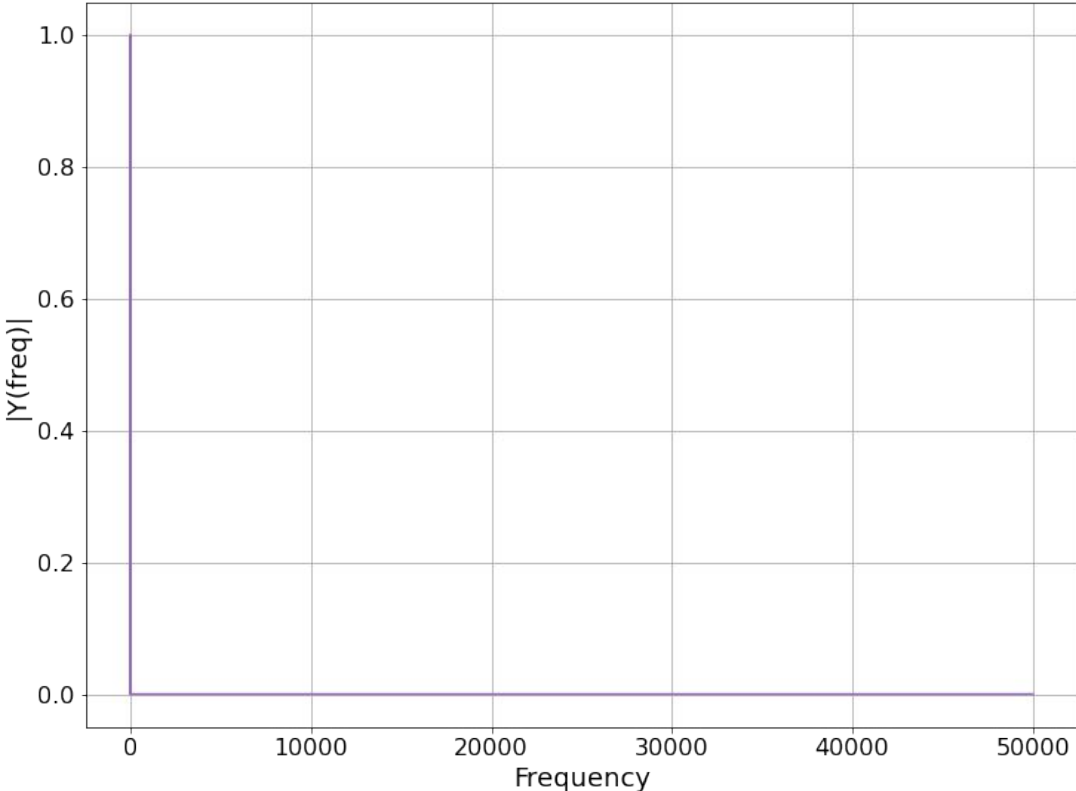
$$\text{dB} = 20 \log_{10} \frac{V_1}{V_2}$$

Why? It actually makes the math easier to do!

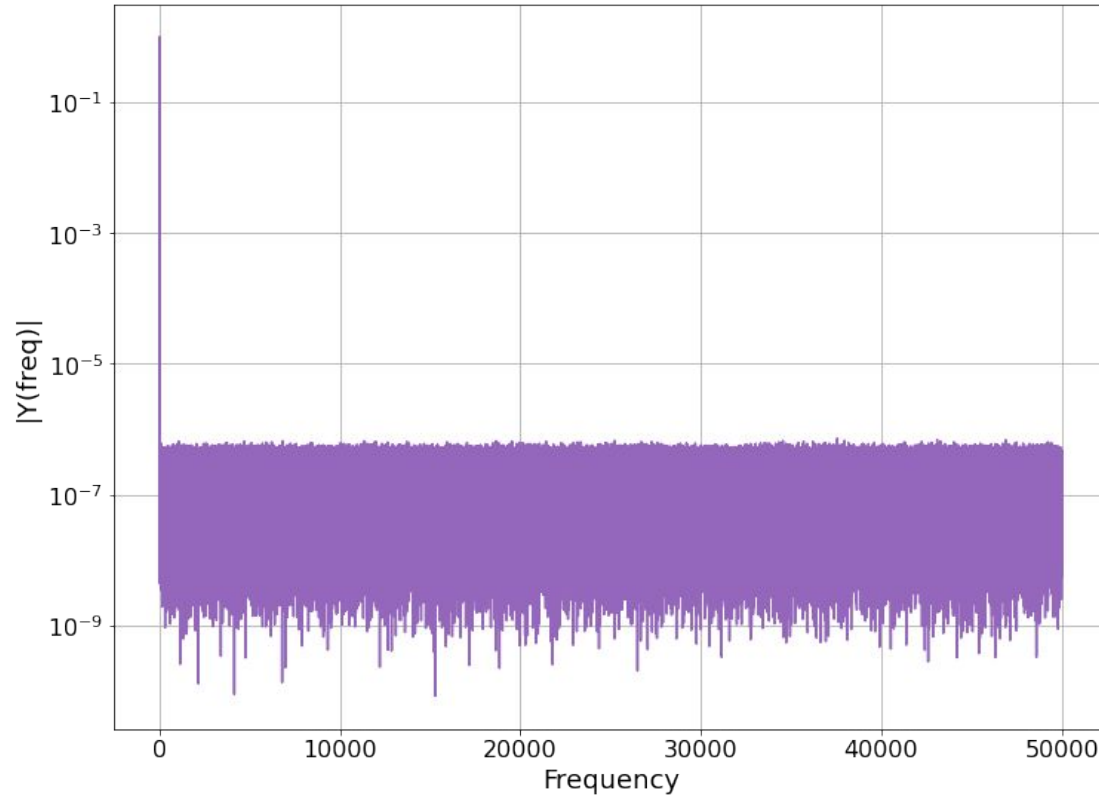
On a scope we are limited in range by the linear scale



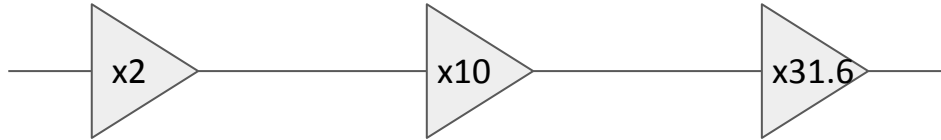
The frequency domain is not any better



But with log scaling we can see details over a much larger range



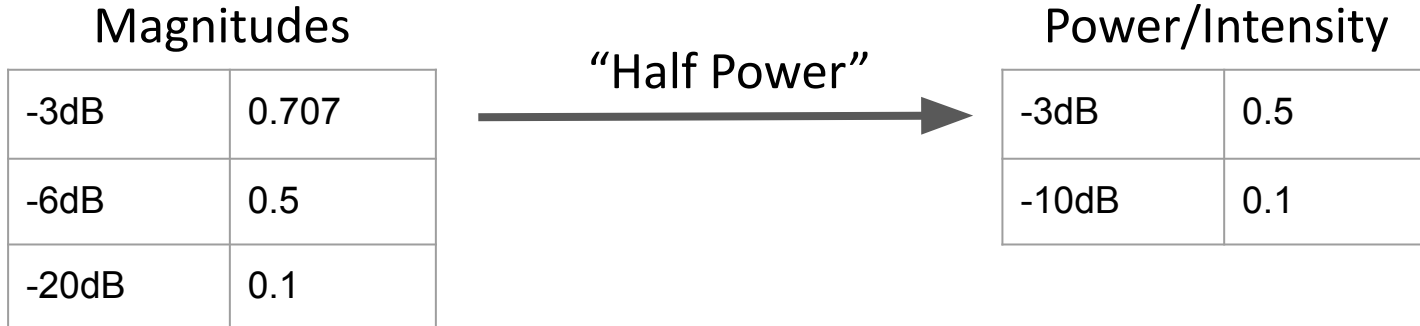
Let's consider a chain of amplifiers and we want to know the total amount of gain applied to an input signal



$$\text{In Gain: } 2 * 10 * 31.6 = 632$$

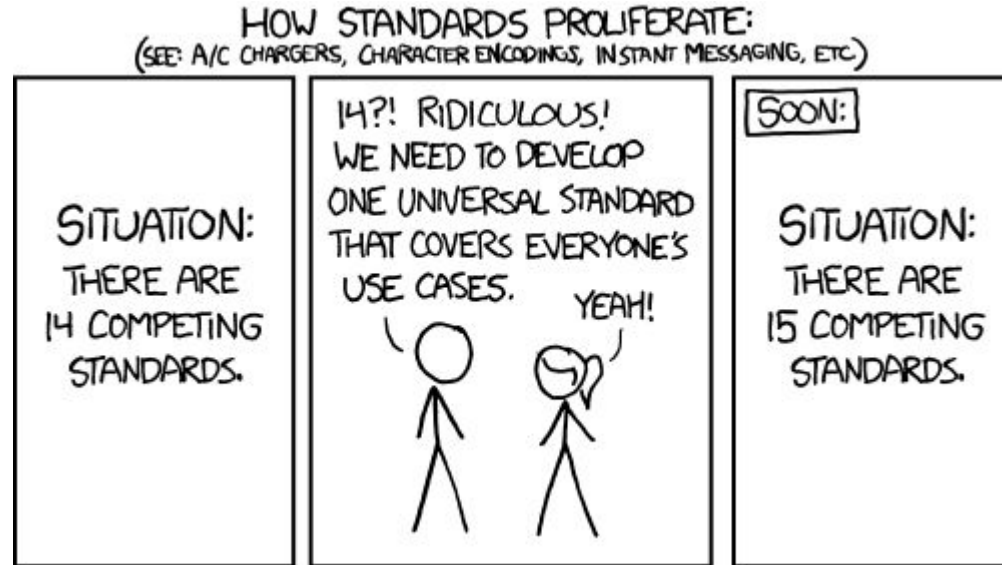
$$\text{In dB: } 6\text{dB} + 20\text{dB} + 30\text{dB} = 56 \text{ dB}$$

Learn the dB “pocket numbers” to make your life easier



There are many industry “standard” dB ratings

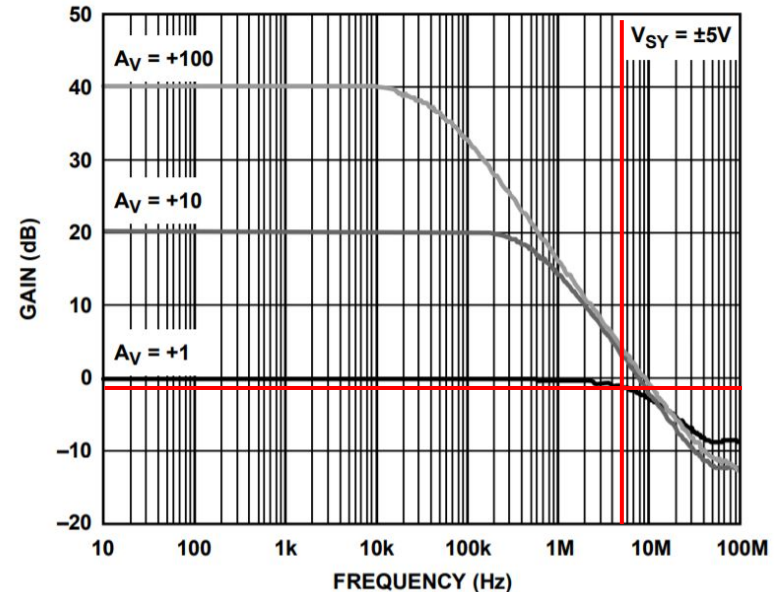
dBm	1 milliwatt
dBu	1 microwatt
dBV	1 Volt
dbmV	1 millivolt
dbmA	1 milliamp



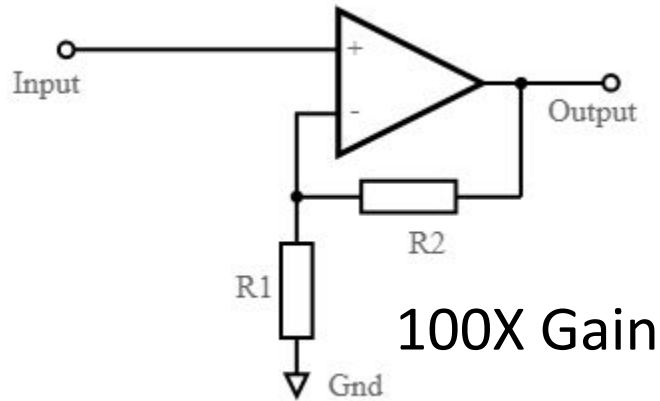
Gain Bandwidth Product (GBWP) is one of the most important considerations

DYNAMIC PERFORMANCE				
Slew Rate	SR	$R_L = 10\text{ k}\Omega, C_L = 100\text{ pF}, A_V = 1$	3.3	V/ μ s
Settling Time	t_s	To 0.1%, $V_{IN} = 10\text{ V step}, C_L = 100\text{ pF}, R_L = 10\text{ k}\Omega, A_V = 1$	3.5	μ s
Gain Bandwidth Product	GBP	$R_L = 10\text{ k}\Omega, A_V = 1$	1.4	MHz
Phase Margin	Φ_M	$R_L = 10\text{ k}\Omega, A_V = 1$		
Channel Separation (ADA4062-2 Only)	CS	$f = 1\text{ kHz}$		
Channel Separation (ADA4062-4 Only)	CS	$f = 1\text{ kHz}$		

$$\text{Bandwidth} = \frac{\text{GBWP}}{\text{Gain}}$$



Let's consider the non-inverting amplifier case

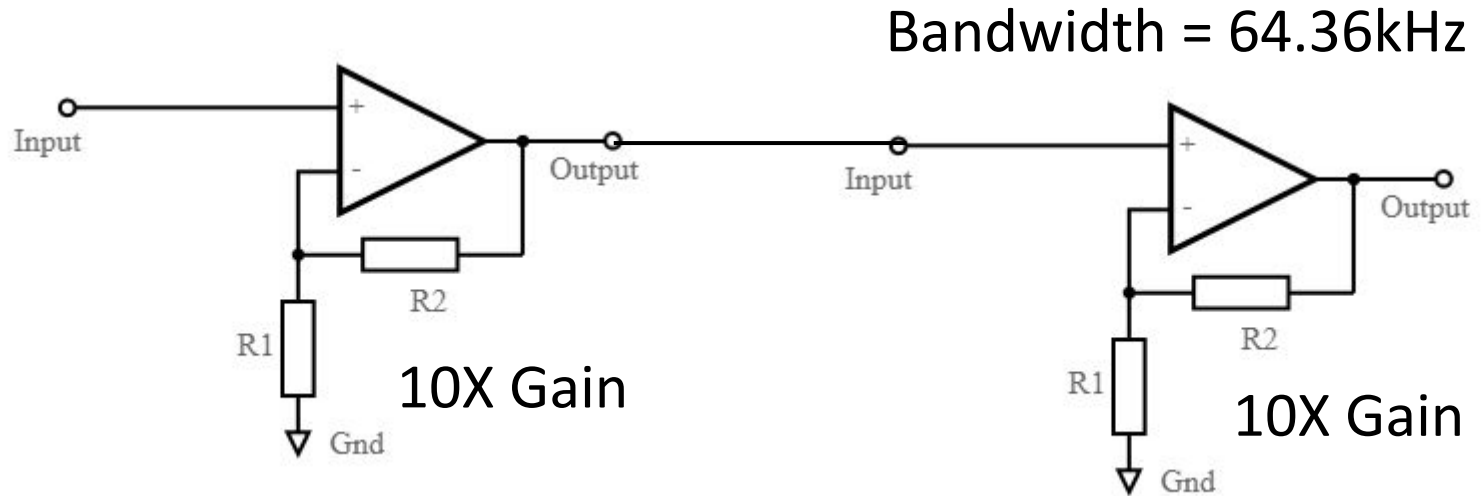


Bandwidth = 10kHz

GBWP = 1MHz

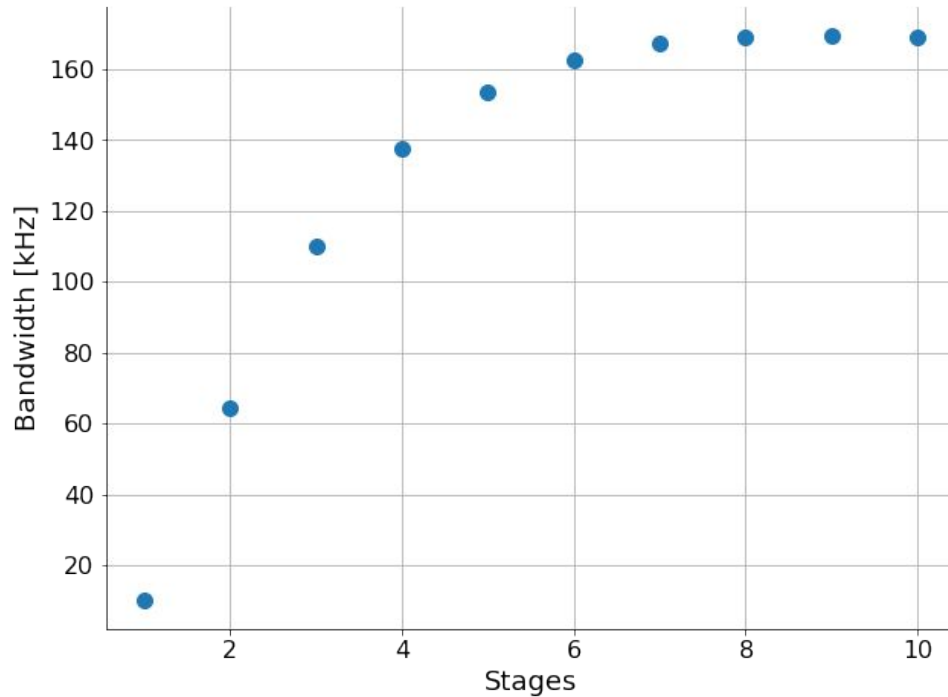
Cascading amplifiers can greatly improve the amplifier bandwidth

$$\text{Bandwidth}_{\text{total}} = \text{Bandwidth} \sqrt{2^{\frac{1}{N}} - 1}$$



GBWP = 1MHz

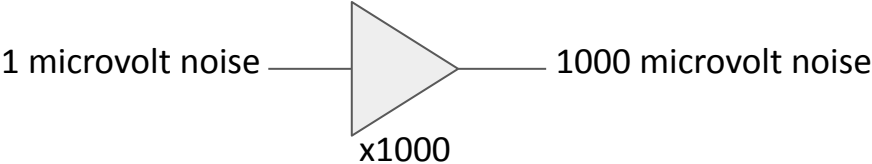
There are decreasing returns with increasing stages



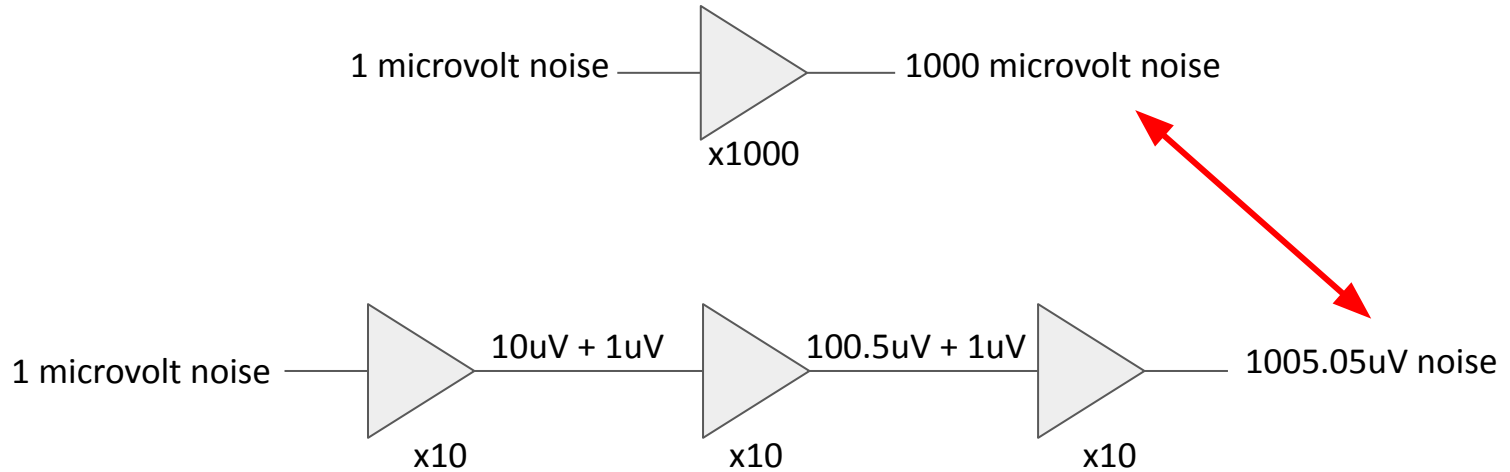
Stages	Stage Gain	BW [kHz]
1	100	10
2	10	64.36
3	4.64	109.83
4	3.16	137.55
5	2.51	153.52
6	2.15	162.43
7	1.93	167.10
8	1.78	169.18
9	1.67	169.62
10	1.58	169.04

$$\text{Bandwidth}_{\text{total}} = \text{Bandwidth} \sqrt{2^{\frac{1}{N}} - 1}$$

So just how bad is the noise amplification from op amp input noise?



So just how bad is the noise amplification from op amp input noise?



$$\text{Noise}_{\text{total}} = \sqrt{N_1^2 + N_2^2 + \dots + N_n^2}$$

Don't forget about ensuring a path for input bias currents

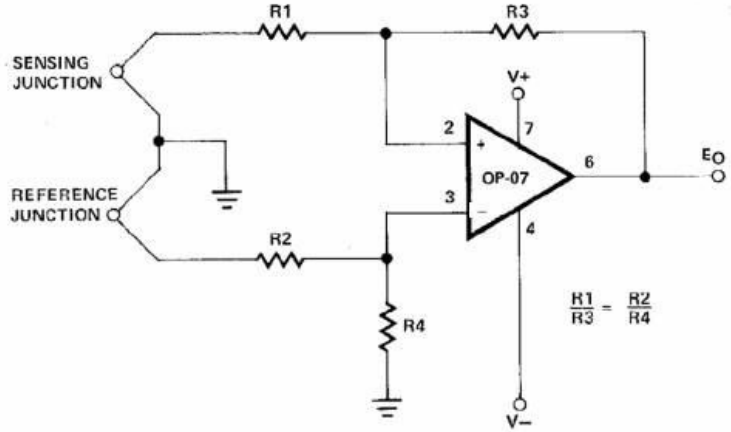
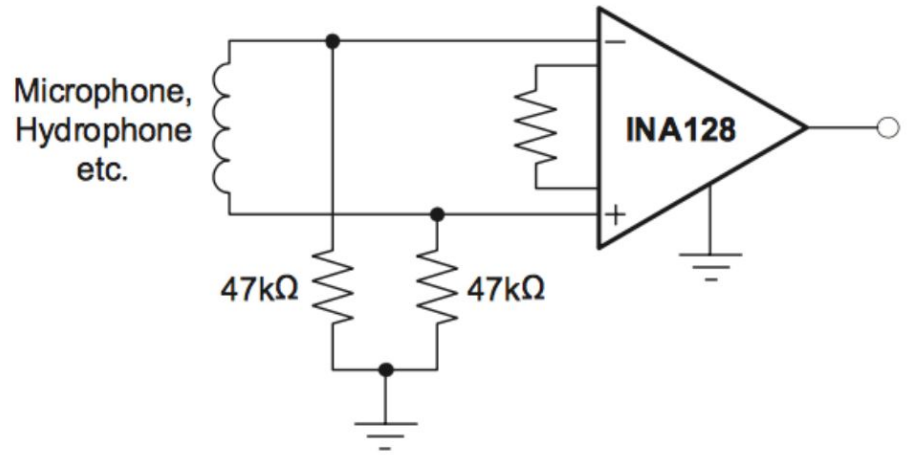
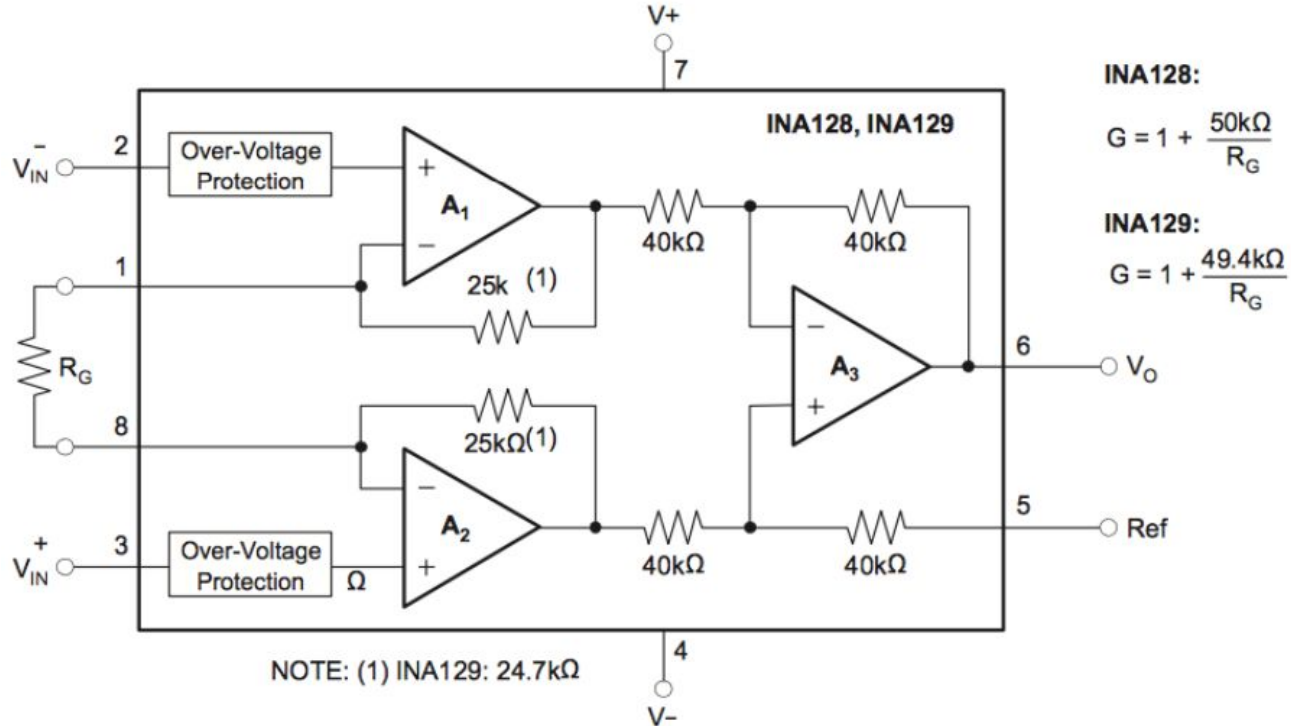


Figure 8. High-Stability Thermocouple Amplifier



Instrumentation amplifiers are precision differential amplifiers

Simplified Schematic



IA's are perfect for bridge circuits/transducers

